

Robust Kalman Filters for Linear Time-Varying Systems with Stochastic Parametric Uncertainties*

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Abstract

We present a robust recursive Kalman filtering algorithm that addresses estimation problems that arise in linear time-varying systems with stochastic parametric uncertainties. The filter has a one-step predictor-corrector structure and minimizes an upper bound of the mean square estimation error at each step, with the minimization reduced to a convex optimization problem based on linear matrix inequalities. The algorithm is shown to converge when the system is mean square stable and the state space matrices are time-invariant. A numerical example, consisting of equalizer design for a communication channel, demonstrates that our algorithm offers considerable improvement in performance when compared to conventional Kalman filtering techniques.

Keywords: Robust Kalman filters, linear time-varying systems, stochastic parametric uncertainty, linear matrix inequality.

1 Introduction

Notations in this paper are fairly standard. $E[\cdot]$ denotes the expectation of a random variable (matrix). $Var(\cdot)$ denotes the variance of a random variable (vector). $P > 0$ ($P \geq 0$) means that P is a symmetric and positive definite (positive semi-definite) matrix. $A \not\geq B$ means that $A \geq B$ and $A \neq B$. $\mathbf{Co}\{\cdot\}$ denotes a convex hull. $\mathbf{diag}(\cdot)$ defines a (block) diagonal matrix. $\|\cdot\|$ is the matrix norm, that is the largest singular value of a matrix. \dagger denotes the Moore-Penrose pseudo inverse of a matrix. $(*)^T$ and $(**)^T$ are used in some places to represent the (2,1) and (3,1) terms of a symmetric matrix when the (1,2) and (1,3) terms are given. $\hat{x}(m|n)$ is the estimation of $x(m)$ with observations up to time n . If $m > n$, $\hat{x}(m|n)$ is known as a predicted estimation. If $m = n$, $\hat{x}(m|n)$ is known as a filtered estimation. For discrete-time systems considered in the paper, the state matrix A is said to be stable if all the eigenvalues of A are strictly inside the unit circle. A discrete-time system $x(k+1) = A(k)x(k)$ where $A(k)$ is a random process, is said to be *mean square stable* ([1]) if for all initial conditions $x(0)$, we have $\lim_{k \rightarrow \infty} E[x(k)x(k)^T] = 0$, i.e., $\lim_{k \rightarrow \infty} x(k) = 0$ almost surely.

Consider the linear system

$$x(k+1) = \mathcal{A}(k)x(k) + \mathcal{B}(k)u_i(k), \quad y(k) = \mathcal{C}(k)x(k) + \mathcal{D}(k)u_m(k), \quad (1)$$

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where $k = 0, 1, 2, \dots$, $x(k) \in \mathbf{R}^n$ is the state, $u_i(k) \in \mathbf{R}^{n_w}$ is the input noise, $y(k) \in \mathbf{R}^{n_y}$ is the measured output, and $u_m(k) \in \mathbf{R}^{n_u}$ is the measurement noise, with u_i and u_m being independent white noise random processes. Equations (1) model systems that are affected by both actuator and sensor noises (u_i and u_m respectively). A fundamental problem associated with such systems is that of state estimation, i.e., the optimal estimation of the state $x(k)$ from the noisy measurements $\{y(i), i = 0, 1, \dots, k\}$; the corresponding state estimate is denoted $\hat{x}(k|k)$. Such estimation problems arise in several applications in signal processing, communications and control; see for example, [2, 3] and the references therein.

Recursive minimum mean-square error (MMSE) estimators form an important class of optimal state estimators for system (1), and have many applications in signal processing, communications and automatic control [3, 4, 5, 6, 7]. MMSE estimators minimize the expected value of the square of the estimation error, i.e., $E[(x(k) - \hat{x}(k|k))^T(x(k) - \hat{x}(k|k))]$, at each k . When the random processes u_i and u_m are Gaussian, it turns out that the MMSE estimator is a linear filter whose coefficients can be determined by solving a Riccati difference equation. (This is the celebrated Kalman filter.) When u_i and u_m are not Gaussian, the Kalman filter yields the best linear MMSE estimator. An important (and desirable) property of the Kalman filtering algorithm is that it converges when system (1) is detectable and stabilizable [8].

The Kalman filter consists of the following two parts¹:

1. One-step prediction update:

$$x_f(k) = \mathcal{A}(k-1)x_f(k-1) + K(k-1)(\mathcal{C}(k-1)x_f(k-1) - y(k-1)), \quad \hat{x}(k|k-1) = x_f(k). \quad (2a)$$

2. Filtered estimation update:

$$\hat{x}(k|k) = \hat{x}(k|k-1) + F(k)(\mathcal{C}(k)\hat{x}(k|k-1) - y(k)). \quad (2b)$$

When the matrices $\mathcal{A}(k)$, $\mathcal{B}(k)$, $\mathcal{C}(k)$, and $\mathcal{D}(k)$ in (1) can be measured exactly, computing the Kalman gains $K(k-1)$ and $F(k)$ in (2) is equivalent to a quadratic optimization problem, one that can be solved analytically [8]. However, in many cases, there exist uncertainties in model parameters and/or model structure because of errors from system identification or model reduction routines; see for example [4, 10, 7, 11, 12]. The performance of estimators designed without accounting for these uncertainties can be severely degraded, and sometimes even unacceptable [13, 12]. Thus, estimators must be designed with graceful performance degradation in the presence of modeling errors. This issue of “robust estimation” has been addressed in a number of recent publications; see for example [14, 12, 15, 16] and the references therein. By assuming the input signal has limited total energy (or ℓ_2 -norm), linear time-invariant (LTI) filters have been designed (see [14, 17]) to optimize the steady state performance measured via the H_∞ norm (or ℓ_2 -gain) of the map from the noise input to the estimation error. In another scenario where the input signal is white noise

¹A more complete introduction to Kalman filter can be found in the literature; see for example [8, 9].

	Input signal	Uncertainty	Design criterion	Filter structure
Robust Kalman	white noise	white noise	MMSE	recursive
Kalman ([8])	white noise	—	MMSE	recursive
Robust H_2 ([15])	white noise	norm bounded	H_2	LTI
Robust H_∞ ([14])	ℓ_2 signal	norm bounded	H_∞	LTI
Stochastic H_∞ ([17])	mean energy bounded	white noise	stochastic H_∞	LTI
Game theoretic estimator ([12])	2-norm bounded	norm bounded	finite horizon quadratic objective	recursive

Table 1: Comparison of several estimation methods.

with limited power, linear time-invariant filters have been designed (see [15]) to optimize the steady state performance measured via the H_2 norm of the map from the noise input to the estimation error.

The models considered in robust estimation problems fall under two classes. One class consists of a linear time-invariant system affected by parametric uncertainties that are deterministic, and typically known only to lie in some bounded set [14, 18, 12]. The second class consists of linear time-invariant systems affected by stochastic uncertainties, which can also be viewed as a multiplicative noise inputs; see [19, 20, 17].

Table 1 summarizes the characteristics of several estimation problems and their solutions. While the above-cited works on robust estimation in the literature provide a fairly complete solution to several steady-state estimation problems, the solutions are linear time-invariant filters, and none of them consider the transient behavior of their estimation algorithms. Indeed, even the conventional Kalman filter is initialized in an *ad hoc* fashion, leaving room for improvement in its transient performance.

In this paper, we consider MMSE estimation problems for linear time varying systems affected by stochastic uncertainties, with a view towards optimizing the transient performance of the estimation. The stochastic uncertainties that we consider affect the system matrices; in addition, we assume that the correlation of the state initial condition is known only to lie in a polytope². For such systems, starting with the standard one-step predictor-corrector filter structure (2), we develop a recursive estimation algorithm where at each step, an upper bound of the mean square of the estimation error over all possible uncertainties is minimized. The minimization is performed via numerical convex optimization over Linear Matrix Inequalities (LMIs). We will refer to our algorithm as the Robust Kalman Filtering algorithm. As a by-product of our robust filtering algorithm, we also obtain a technique for optimally initializing a recursive filtering algorithm, for instance the conventional Kalman filter.

We will demonstrate through an example that the robust Kalman filter can provide much improved transient performance when compared to the conventional Kalman filter. Perhaps more important, for systems with stochastic parametric uncertainties, the performance of the conven-

²Our framework is perhaps related closest with the one in [10], where the uncertainties affect the noise moments; a game-theoretic argument is used to establish the existence of an optimal recursive scheme.

tional Kalman filter can be severely degraded, while the performance of the robust Kalman filter degrades fairly gracefully.

As with the conventional Kalman filtering algorithm for time-varying systems, the convergence of the robust Kalman filtering algorithm that we present is not guaranteed in general. However, we prove convergence in the special case of the estimation of a system with time-invariant state space matrices and stochastic parametric uncertainties, provided that the uncertain system is mean-square stable. Moreover, we show that the conventional Kalman filter is a special case of the proposed robust Kalman filtering algorithm for systems with no uncertainties.

The organization of the paper is as follows: In Section 2, we discuss the mathematical framework underlying our problem, and make some preliminary remarks. In Section 3, we describe the derivation of a robust Kalman filtering algorithm that minimizes an upper bound of the mean square of the estimation error at each step. We also present the convergence property of this recursive algorithm, and its connection with the conventional Kalman filter. In Section 4, we apply the filtering technique developed in this paper to design equalizers for a communication channel. The proofs are given in Appendix A.

2 Problem setup

We consider the following linear time-varying system³:

$$x(k+1) = A_{\Delta}(k)x(k) + B_{\Delta}(k)w(k), \quad y(k) = C_{\Delta}(k)x(k) + D_{\Delta}(k)w(k), \quad z(k) = L(k)x(k), \quad (3a)$$

where

$$\begin{aligned} A_{\Delta}(k) &= A(k) + \sum_{i=1}^m A_i^s(k)\zeta_i^s(k), & B_{\Delta}(k) &= B(k) + \sum_{i=1}^m B_i^s(k)\zeta_i^s(k), \\ C_{\Delta}(k) &= C(k) + \sum_{i=1}^m C_i^s(k)\zeta_i^s(k), & D_{\Delta}(k) &= D(k) + \sum_{i=1}^m D_i^s(k)\zeta_i^s(k), \end{aligned} \quad (3b)$$

$x(k) \in \mathbf{R}^n$, $y(k) \in \mathbf{R}^{n_y}$, and $w(k) \in \mathbf{R}^{n_w}$. $z(k) \in \mathbf{R}^{n_z}$ is the signal we wish to estimate. w is a zero-mean white noise process, and satisfies $E[w(i)w(j)^T] = \delta(i-j)I$, where $\delta(k)$ is the Dirac Delta function. ζ_i^s , $i = 1, \dots, m$ are zero-mean random processes with $E[\zeta_i^s(k)\zeta_j^s(l)] = \delta(i-j)\delta(k-l)$. The initial state $x(0)$ of system (3a) is a random vector, with its correlation $X(0) = E[x(0)x(0)^T]$ known only to lie in a polytope $\mathbf{Co}\{X_1(0), \dots, X_p(0)\}$. The random processes w , ζ_i^s , $i = 1, \dots, m$, and the random vector $x(0)$ are mutually independent.

System (3) is said to be *mean square stable* ([1]) if with $w(k) = 0$ for $k = 0, 1, 2, \dots$, we have

$$\lim_{k \rightarrow \infty} E[x(k)x(k)^T] = 0$$

for any initial condition $x(0)$.

Our objective is to design an optimal robust filter of the one-step predictor-corrector form given in (2) with $\hat{z}(k) = L(k)\hat{x}(k|k)$ being an estimate of $z(k)$ (see Figure 1). (The case $L(k) = I$

³ $w(k)$ is a vector containing both input noise and measurement noise.

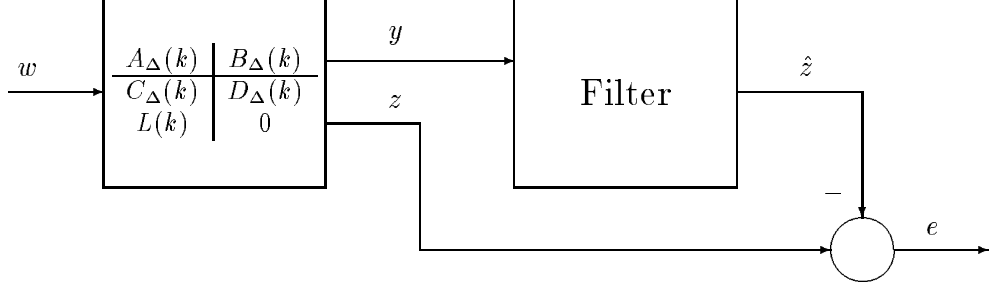


Figure 1: Estimation of linear time-varying systems with stochastic parametric uncertainties

corresponds to state estimation.) Specifically, since the correlation $X(k)$ of the state at each k depends on the correlation $X(0)$ of the initial condition, $X(k)$ is uncertain when $X(0)$ is uncertain. We wish to find the optimal Kalman gains $K(k-1)$ and $F(k)$ so as to minimize the maximum value of the mean square of the estimation error, $E[\|z(k) - \hat{z}(k)\|^2]$, over all allowable values for $X(k)$. (Here, the expectation is taken with respect to the random initial state, the input and measurement noises, as well as the stochastic parametric uncertainties ζ_i^s .)

Compared to the steady state H_2 and H_∞ filters in [14, 15, 17], the filters we design are recursive and optimize the transient performance. In addition, the filters we design are robust to the stochastic parametric uncertainties, in comparison to the conventional (non-robust) Kalman filters [8, 9], which are designed based on nominal models with no uncertainties.

3 Robust Kalman filtering

3.1 System model

We begin by rewriting equations (3) in an equivalent form:

$$x(k+1) = A(k)x(k) + v_x(k), \quad y(k) = C(k)x(k) + v_y(k), \quad z(k) = L(k)x(k),$$

where

$$v_x(k) = \sum_{i=1}^m A_i^s(k) \zeta_i^s(k) x(k) + \left(B(k) + \sum_{i=1}^m B_i^s(k) \zeta_i^s(k) \right) w(k)$$

and

$$v_y(k) = \sum_{i=1}^m C_i^s(k) \zeta_i^s(k) x(k) + \left(D(k) + \sum_{i=1}^m D_i^s(k) \zeta_i^s(k) \right) w(k).$$

v_x and v_y are random processes. Since ζ_i^s , $i = 1, \dots, m$ and w are independent zero-mean white-noise processes, the mean and variance of v_x and v_y satisfy the following conditions for every $k = 0, 1, 2, \dots$

$$E[v_x(k)] = 0, \quad E[v_y(k)] = 0, \\ \text{Var} \left(\begin{bmatrix} v_x(k) \\ v_y(k) \end{bmatrix} \right) = \begin{bmatrix} \hat{B}(k) \hat{B}(k)^T & \hat{B}(k) \hat{D}(k)^T \\ \hat{D}(k) \hat{B}(k)^T & \hat{D}(k) \hat{D}(k)^T \end{bmatrix}.$$

where

$$\begin{aligned}\hat{B}(k) &= \begin{bmatrix} A_1^s(k)X(k)^{1/2} & \dots & A_m^s(k)X(k)^{1/2} & B(k) & B_1^s(k) & \dots & B_m^s(k) \end{bmatrix}, \\ \hat{D}(k) &= \begin{bmatrix} C_1^s(k)X(k)^{1/2} & \dots & C_m^s(k)X(k)^{1/2} & D(k) & D_1^s(k) & \dots & D_m^s(k) \end{bmatrix},\end{aligned}$$

and $X(k) = E[x(k)x(k)^T]$. We also note that v_x and v_y are white-noise random processes, so that when $i \neq j$, $E[v_x(i)v_x(j)^T] = 0$, $E[v_y(i)v_y(j)^T] = 0$ and $E[v_x(i)v_y(j)^T] = 0$. In addition, $E[x(i)v_x(j)^T] = 0$ and $E[x(i)v_y(j)^T] = 0$ for $i \leq j$. Therefore, random processes $[v_x, v_y]$ and $[\hat{B}(k)v(k), \hat{D}(k)v(k)]$ have the same (first- and second-order) statistics, where v is a zero-mean unit-variance white noise random process.

Thus the model in (3) can be rewritten as

$$x(k+1) = A(k)x(k) + \hat{B}(k)v(k), \quad y(k) = C(k)x(k) + \hat{D}(k)v(k), \quad z(k) = L(k)x(k). \quad (4)$$

If the variance $X(0) = E[x(0)x(0)^T]$ is known, then it is easily verified that $X(k)$ is uniquely determined by the following recursion:

$$X(k+1) = h(X(k)) \triangleq A(k)X(k)A(k)^T + B(k)^T B(k) + \sum_{j=1}^m \left(A_j^s(k)X(k)A_j^s(k)^T + B_j^s(k)B_j^s(k)^T \right). \quad (5)$$

However, in our setting, $X(0)$ is only known to lie in a polytope $\mathbf{Co}\{X_1(0), \dots, X_p(0)\}$. Since in the recursion (5), $X(k+1)$ is a linear function of $X(k)$, it can be shown recursively that $X(k)$ also lies in a polytope $\mathbf{Co}\{X_1(k), \dots, X_p(k)\}$, where

$$X_i(k+1) = A(k)X_i(k)A(k)^T + B(k)B(k)^T + \sum_{j=1}^m \left(A_j^s(k)X_i(k)A_j^s(k)^T + B_j^s(k)B_j^s(k)^T \right). \quad (6)$$

If the system is mean square stable and the state space matrices are time-invariant, the polytope converges to a fixed point as $k \rightarrow \infty$; see Proposition A.3 in Appendix A.

3.2 Robust Kalman Filter Recursions

While (4) is similar to the setting for the conventional Kalman filtering problem, we note that the system matrices $\hat{B}(k)$ and $\hat{D}(k)$ depend on the second moment of the states, and are thus uncertain. For this reason, the optimal linear recursive MMSE filter design problem is considerably harder than the conventional Kalman filtering problem, which can be solved analytically [8].

With uncertain system (3), at each k , we are given:

- (N1) $L(k)$ and $A(k)$, $B(k)$, $C(k)$, $D(k)$, which are the measurements of the system matrices $A_\Delta(k)$, $B_\Delta(k)$, $C_\Delta(k)$, $D_\Delta(k)$. Note that the measurement noise in system matrices is characterized by the terms of ζ_i^s in (3b).
- (N2) $y(k)$, which is the measurement of the noisy output. Note that the measurement noise in output is characterized by the terms of w in (3a).

(N3) From (6), the vertices $X_1(k), \dots, X_p(k)$ of the polytope where $X(k)$ lies in.

Our objective is to design a linear MMSE estimator:

$$x_f(k) = A(k-1)x_f(k-1) + K(k-1)(C(k-1)x_f(k-1) - y(k-1)), \quad \hat{x}(k|k-1) = x_f(k), \quad (7a)$$

$$\hat{x}(k|k) = \hat{x}(k|k-1) + F(k)(C(k)\hat{x}(k|k-1) - y(k)), \quad (7b)$$

where the Kalman gains $K(k-1)$ and $F(k)$ are obtained from the solution of the following minimax optimization problem:

$$\begin{aligned} \text{Minimize:} \quad & \max_{X(k)} E \left[\|z(k) - \hat{z}(k)\|^2 \right] \\ \text{Subject to:} \quad & \text{Data (N1-N3),} \\ & \text{Equations (7) and (4).} \end{aligned} \quad (8)$$

The minimax problem (8) is difficult to solve directly. We therefore first define an upper bound on the quantity

$$J(K(k-1), F(k)) \triangleq \max_{X(k)} E \left[\|z(k) - \hat{z}(k)\|^2 \right],$$

and then determine $K(k-1)$ and $F(k)$ to minimize the upper bound. We will show that this results in a convex optimization problem.

We now proceed with deriving an upper bound on $J(K(k-1), F(k))$. The estimation algorithm we will derive is a recursive algorithm. The initialization of the recursions will be discussed in Section 3.3. At this stage, we focus our attention on the recursions. We assume that at each k , we have available a matrix $P(k-1)$ with

$$P(k-1) \geq E \left[(x(k-1) - \hat{x}(k-1|k-2))(x(k-1) - \hat{x}(k-1|k-2))^T \right].$$

Then,

$$\begin{aligned} & E \left[(x(k) - \hat{x}(k|k-1))(x(k) - \hat{x}(k|k-1))^T \right] \\ & \leq \sup_{X(k-1)} (A(k-1) + K(k-1)C(k-1))P(k-1)(A(k-1) + K(k-1)C(k-1))^T \\ & \quad + (\hat{B}(k-1) + K(k-1)\hat{D}(k-1))(\hat{B}(k-1) + K(k-1)\hat{D}(k-1))^T \end{aligned} \quad (9)$$

$$\triangleq f(P(k-1), K(k-1)), \quad (10)$$

where $X(k-1)$ lies in $\mathbf{Co}\{X_1(k-1), \dots, X_p(k-1)\}$.

Then, we can obtain $P(k) \geq E \left[(x(k) - \hat{x}(k|k-1))(x(k) - \hat{x}(k|k-1))^T \right]$, simply by requiring that

$$f(P(k-1), K(k-1)) \leq P(k). \quad (11)$$

Next, we have

$$E \left[(x(k) - \hat{x}(k|k))(x(k) - \hat{x}(k|k))^T \right] \leq \sup_{X(k)=h(\bar{X}(k-1))} (I + F(k)C(k))P(k)(I + F(k)C(k))^T + (F(k)\hat{D}(k))(F(k)\hat{D}(k))^T \quad (12)$$

$$\triangleq \tau(F(k), P(k)), \quad (13)$$

where $X(k-1)$ lies in $\mathbf{Co}\{X_1(k-1), \dots, X_p(k-1)\}$. Then, for any matrix $M(k)$ that satisfies

$$\tau(F(k), P(k)) \leq M(k), \quad (14)$$

we have

$$E \left[(z(k) - \hat{z}(k))(z(k) - \hat{z}(k))^T \right] \leq L(k)M(k)L(k)^T,$$

which implies that $\mathbf{Tr}(L(k)M(k)L(k)^T)$ is an upper bound of the objective function in (8).

We can now formulate the problem of determining $K(k-1)$ and $F(k)$ so as to minimize an upper bound on the the mean square of the estimation error over all possible values for the corresponding state correlation matrix as follows:

$$\begin{aligned} \text{Minimize:} & \quad \mathbf{Tr}(L(k)M(k)L(k)^T) \\ & \quad P(k), M(k), K(k-1), F(k) \\ \text{Subject to:} & \quad \text{Conditions (11) and (14),} \end{aligned} \quad (15)$$

where $P(k)$, $M(k)$, $K(k-1)$ and $F(k)$ are optimization variables.

While the optimization problem (15) has no analytical solution in general, we establish via the following theorem that it can be reformulated as a convex optimization problem, that of minimizing a linear objective subject with Linear Matrix Inequality (LMI) constraints. This problem can be solved numerically very efficiently using standard algorithms [21, 22], so that Theorem 3.1 provides for an efficient and effective numerical solution of problem (15); for details on LMIs, we refer readers to the book [1] and the references therein.

Theorem 3.1 *Consider the following optimization problem of minimizing a linear objective, subject to LMI constraints, in the variables $Q(k)$, $M(k)$, $Y(k)$, and $F(k)$.*

$$\begin{aligned} \text{Minimize:} & \quad \mathbf{Tr}(L(k)M(k)L(k)^T) \\ & \quad Q(k), M(k), Y(k), F(k) \end{aligned} \quad (16a)$$

Subject to:

$$\begin{bmatrix} Q(k) & T_{12}(k-1) & T_{13,i}(k-1) \\ T_{12}(k-1)^T & Q(k-1) & 0 \\ T_{13,i}(k-1)^T & 0 & I \end{bmatrix} \geq 0, \quad (16b)$$

$$\begin{bmatrix} M(k) & I + F(k)C(k) & F(k)\hat{D}_i(k) \\ I + C(k)^T F(k)^T & Q(k) & 0 \\ \hat{D}_i(k)^T F(k)^T & 0 & I \end{bmatrix} \geq 0, \quad i = 1, \dots, p, \quad (16c)$$

$$Q(k) \geq 0, \quad (16d)$$

where

$$\begin{aligned}
T_{12}(k-1) &= Q(k)A(k-1) + Y(k)C(k-1), \\
T_{13,i}(k-1) &= Q(k)\hat{B}_i(k-1) + Y(k)\hat{D}_i(k-1), \\
\hat{B}_i(k) &= \begin{bmatrix} A_1^s(k)X_i(k)^{1/2} & \dots & A_m^s(k)X_i(k)^{1/2} & B(k) & B_1^s(k) & \dots & B_m^s(k) \end{bmatrix}, \\
\hat{D}_i(k) &= \begin{bmatrix} C_1^s(k)X_i(k)^{1/2} & \dots & C_m^s(k)X_i(k)^{1/2} & D(k) & D_1^s(k) & \dots & D_m^s(k) \end{bmatrix},
\end{aligned} \tag{16e}$$

and $X_i(k)$ is defined in (6). Then:

1. Suppose that $P(k)$, $M(k)$, $K(k-1)$ and $F(k)$ satisfy conditions (11) and (14), then $Q(k) = P(k)^\dagger$, $M(k)$, $Y(k) = P(k)^\dagger K(k-1)$ and $F(k)$ satisfy conditions (16b), (16c) and (16d).
2. Suppose that $Q(k)$, $M(k)$, $Y(k)$ and $F(k)$ satisfy conditions (16b), (16c) and (16d), then $P(k) = Q(k)^\dagger$, $M(k)$, $K(k-1) = Q(k)^\dagger Y(k)$ and $F(k)$ satisfy conditions (11) and (14).
3. The optimal value of Problem (16) equals the optimal value of Problem (15).

Proof: Let $X(k-1) = \sum_{i=1}^p \lambda_i X_i(k-1)$, where $\lambda_i \in [0, 1]$ and $\sum_{i=1}^p \lambda_i = 1$. Since $h(X(k-1))$ (Eq. (5)) is a linear function in $X(k-1)$, we have $X(k) = \sum_{i=1}^p \lambda_i X_i(k)$ for the same set of λ_i , $i = 1, \dots, p$. By Schur's complements lemma, condition (11) is equivalent to

$$\sum_{i=1}^p \lambda_i \begin{bmatrix} P(k) & \hat{T}_{12}(k) & \hat{T}_{13}(k)\mathcal{X}_i(k-1) \\ (*)^T & P(k-1) & 0 \\ (**)^T & 0 & \mathcal{X}_i^2(k-1) \end{bmatrix} \geq 0, \tag{17}$$

where $(*)$ is the (1, 2) term of the symmetric matrix, and $(**)$ is the (1, 3) term of the symmetric matrix,

$$\begin{aligned}
\hat{T}_{12}(k) &= (A(k-1) + K(k-1)C(k-1))P(k-1), \\
\hat{T}_{13}(k) &= \hat{B}_i(k-1) + K(k-1)\hat{D}_i(k-1).
\end{aligned}$$

Similarly, condition (14) is equivalent to

$$\sum_{i=1}^p \lambda_i \begin{bmatrix} M(k) & \tilde{T}_{12}(k) & \tilde{T}_{13}(k)\mathcal{X}_i(k) \\ (*)^T & P(k) & 0 \\ (**)^T & 0 & \mathcal{X}_i^2(k) \end{bmatrix} \geq 0, \tag{18}$$

where $\mathcal{X}_i(k) = \mathbf{diag}(X_i(k)^{1/2}, \dots, X_i(k)^{1/2}, I, \dots, I)$ and

$$\tilde{T}_{12}(k) = (I + F(k)C(k))P(k), \quad \tilde{T}_{13}(k) = F(k)\hat{D}_i(k).$$

Since conditions (17) and (18) hold for all $\lambda_i \in [0, 1]$ such that $\sum_{i=1}^p \lambda_i = 1$, they are equivalent respectively to

$$\begin{bmatrix} P(k) & \hat{T}_{12}(k) & \hat{T}_{13}(k) \\ (*)^T & P(k-1) & 0 \\ (**)^T & 0 & I \end{bmatrix} \geq 0, \tag{19}$$

and

$$\begin{bmatrix} M(k) & (I + F(k)C(k))P(k) & F(k)\hat{D}_i(k) \\ (*)^T & P(k) & 0 \\ (**)^T & 0 & I \end{bmatrix} \geq 0, \quad (20)$$

Item (1): Suppose that $P(k)$, $M(k)$, $K(k-1)$ and $F(k)$ satisfy conditions (11) and (14), and thus satisfy LMIs (19) and (20). Then, multiplying LMI (19) on the left and right by $\mathbf{diag}(P(k)^\dagger, P(k-1)^\dagger, I)$ yields LMI (16b); multiplying LMI (20) on the left and right by $\mathbf{diag}(I, P(k)^\dagger, I)$ yields LMI (16c), where $Q(k) = P(k)^\dagger$ and $Y(k) = P(k)^\dagger K(k-1)$. In addition, $P(k) \geq 0$ implies $Q(k) \geq 0$.

Item (2): Following the same line as in item (1), multiplying LMI (16b) on the left and right by $\mathbf{diag}(Q(k)^\dagger, Q(k-1)^\dagger, I)$ yields LMI (19); multiplying LMI (16c) on the left and right by $\mathbf{diag}(I, Q(k)^\dagger, I)$ yields LMI (20), where $P(k) = Q(k)^\dagger$ and $K(k-1) = Q(k)^\dagger Y(k)$. Therefore the result of item (2) is established.

Item (3): Item (1) implies that the optimal value of Problem (16) is less than or equal to the optimal value of Problem (15); Item (2) implies that the optimal value of Problem (15) is less than or equal to the optimal value of Problem (16). Therefore, we established that the optimal values of the Problem (16) and the Problem (15) are equal. \square

3.3 Initialization of the robust Kalman filter

Theorem 3.1 paves the way for a robust Kalman filtering algorithm. To start the algorithm, we need to initialize $Q(0)$, the process that we describe next. Let $x_f(0) = 0$.⁴ Then we have $Q(0) = E[x(0)x(0)^T]^\dagger = X(0)^\dagger$, where $X(0) \in \mathbf{Co}\{X_1(0), \dots, X_p(0)\}$. Using Theorem 3.1, $Q(1)$ can be computed as an optimizer to the following problem:

$$\begin{aligned} & \underset{Q(1), M(1), Y(1), F(1)}{\text{Minimize:}} && \text{Tr}(L(1)M(1)L(1)^T) \\ & \text{Subject to:} && \\ & && \begin{bmatrix} Q(1) & T_{12}(0) & T_{13,i}(0) \\ T_{12}(0)^T & X(0)^\dagger & 0 \\ T_{13,i}(0)^T & 0 & I \end{bmatrix} \geq 0, \\ & && \begin{bmatrix} M(1) & I + F(1)C(1) & F(1)\hat{D}_i(1) \\ I + C(1)^T F(1)^T & Q(1) & 0 \\ \hat{D}_i(1)^T F(1)^T & 0 & I \end{bmatrix} \geq 0, \\ & && Q(1) \geq 0 \end{aligned} \quad (21)$$

$$X(0) \in \mathbf{Co}\{X_1(0), \dots, X_p(0)\}, \quad i = 1, \dots, p,$$

where $M(1) = M(1)^T$, $Q(1) = Q(1)^T$. We then have $P(1) = Q(1)^\dagger$ and $K(0) = Q(1)^\dagger Y(1)$.

The matrix inequalities in (21) are not linear in all the variables, because the first LMI constraint has the term $X(0)^\dagger$. However, following the same line as in the proof of Theorem 3.1, it is easily

⁴If $E[x(0)] = a$ is known and $a \neq 0$, we may define $x_f(0) = a$ and $Q(0)$ becomes $(X(0) - aa^T)^\dagger$. The initialization process (22) can still be applied by replacing $X(0)$ by $X(0) - aa^T$ and $X_i(0)$ by $X_i(0) - aa^T$ respectively.

shown that (21) is equivalent to the LMI

$$\begin{aligned}
& \underset{Q(1), M(1), Y(1), F(1)}{\text{Minimize:}} && \text{Tr}(L(1)M(1)L(1)^T) \\
& \text{Subject to:} && \\
& && \begin{bmatrix} Q(1) & T_{12}(0)X_i(0) & T_{13,i}(0) \\ X_i(0)T_{12}^T(0) & X_i(0) & 0 \\ T_{13,i}(0)^T & 0 & I \end{bmatrix} \geq 0, \quad i = 1, \dots, p, \\
& && \begin{bmatrix} M(1) & I + F(1)C(1) & F(1)\hat{D}_i(1) \\ I + C(1)^T F(1)^T & Q(1) & 0 \\ \hat{D}_i(1)^T F(1)^T & 0 & I \end{bmatrix} \geq 0, \\
& && Q(1) \geq 0,
\end{aligned} \tag{22}$$

where $T_{12}(0)$, $T_{13,i}(0)$, $\hat{D}_i(1)$ are defined in (16e), and $K(0) = Q(1)^\dagger Y(1)$.

Remark: The solution to the optimization problem (21) can also be applied towards optimally initializing other recursive algorithms, for instance the conventional Kalman filtering algorithms, with an attendant improvement in transient performance. \diamond

3.4 Robust Kalman filtering algorithm

We now summarize the various steps in our robust Kalman filtering algorithm:

Robust Kalman filtering algorithm:

Step 1. Solve (22) to initialize $Q(1)$, $M(1)$, $K(0)$ and $F(1)$. Let $k = 1$.

Step 2. At time $k + 1$, let

$$\begin{aligned}
X_i(k+1) = & A(k)X_i(k)A(k)^T + B(k)B(k)^T \\
& + \sum_{j=1}^m \left(A_j^s(k)X_i(k)A_j^s(k)^T + B_j^s(k)B_j^s(k)^T \right), \quad i = 1, \dots, p.
\end{aligned} \tag{23}$$

Step 3. Solve the optimization problem (16) for $Q(k)$, $M(k)$ and $K(k-1)$, $F(k)$.

Step 4. Repeat Steps 2 and 3.

We note that if we have $x_f(0) = E[x(0)]$ in the estimator (7), then the robust Kalman filtering algorithm is unbiased.

In the following, we will discuss the connection of the robust Kalman filter with the conventional Kalman filter. We will also provide the condition for the convergence of the recursions in the robust Kalman filtering algorithm.

3.5 Connection with the conventional Kalman filtering algorithm

If there is no stochastic parametric uncertainty in system (3), i.e., $\zeta_i^s(k) = 0$, it can be shown that the robust Kalman filter reduces to the conventional Kalman filter. Therefore the robust Kalman filter can be viewed as an extension of the conventional Kalman filter to systems with stochastic parametric uncertainties.

Consider the following optimization problem of the conventional Kalman filtering algorithm (see [8] for details):

$$P^*(k+1) = \min_{K(k)} f(P^*(k), K(k)),^5 \quad \text{and} \quad K^*(k) = \operatorname{argmin}_{K(k)} f(P^*(k), K(k)), \quad (24a)$$

where

$$f(P(k), K(k)) = (A + K(k)C)P(k)(A + K(k)C)^T + (K(k)D + B)(K(k)D + B)^T, \quad (24b)$$

(A, B, C, D) are the state space matrices of the system. The optimal solution to Problem (24a) is

$$\begin{aligned} K^*(k) &= -(AP^*(k)C^T + BD^T)(CP^*(k)C^T + DD^T)^{-1}, \\ P^*(k+1) &= (A + K^*(k)C)P^*(k)(A + K^*(k)C)^T + (K^*(k)D + B)(K^*(k)D + B)^T. \end{aligned} \quad (24c)$$

Here we assume DD^T is nonsingular, which is a standard assumption in Kalman filtering; see for example [8, 2, 3]. The well-known condition for the convergence of the recursions in (24) is given by Proposition A.2 in Appendix A.

The recursions in the robust Kalman filter are based on solving an optimization problem (16), or equivalently problem (15), at each step. In order to show the equivalence between the robust and the conventional Kalman filtering algorithms under the condition $\zeta_i^s(k) = 0$, it suffices to show that the optimal solution to (24) is the same as the optimal solution to (15), in which $\hat{B}(k) = B$, $\hat{D}(k) = D$.

Theorem 3.2 *Let $P(0) \geq 0$ be given. Let $\{P^*(k), K^*(k)\}$ be the sequence consisting of the solutions to the optimization problem (24). Let $\{\tilde{P}^*(k), \tilde{M}^*(k), \tilde{K}^*(k), \tilde{F}^*(k)\}$ be the sequence consisting of the solutions to the optimization problem*

$$\min_{P(k), M(k), K(k-1), F(k)} \mathbf{Tr}(M(k)), \quad k = 1, 2, \dots, \quad (25a)$$

where

$$\begin{aligned} f(P(k-1), K(k-1)) &= (A + K(k-1)C)P(k-1)(A + K(k-1)C)^T \\ &\quad + (K(k-1)D + B)(K(k-1)D + B)^T, \end{aligned} \quad (25b)$$

$$P(k) = f(P(k-1), K(k-1)), \quad (25c)$$

$$M(k) = (I + F(k)C)P(k)(I + F(k)C)^T + F(k)DD^T F(k)^T, \quad (25d)$$

and DD^T is assumed to be positive definite.

Then:

$$P^*(k) = \tilde{P}^*(k) \quad \text{and} \quad K^*(k) = \tilde{K}^*(k), \quad k = 1, 2, \dots$$

⁵Though the set of positive semidefinite matrices is only a partially ordered set, it is well-known that the problem $\min_{K(k)} f(P^*(k), K(k))$ is well defined; see [8].

We note that Problem (25) is a special case of Problem (15) with $\hat{B}(k) = B$, $\hat{D}(k) = D$. We also note that if the system is time-varying, the equivalence between the robust and the conventional Kalman filters still holds, by replacing (A, B, C, D) with $(A(k), B(k), C(k), D(k))$ in Theorem 3.2. The proof of Theorem 3.2 is given in Appendix A.

As a simple corollary of Theorem 3.2, we have the following conclusion: If the correlation matrix $X(0)$ is exactly known in advance, the polytope covering $X(k)$ is a fixed point at each k ; see the recursion (5). The optimization problem (16) turns out to be equivalent to the optimization problem encountered in the conventional Kalman filtering problem, with B replaced by $\hat{B}(k)$ and D replaced by $\hat{D}(k)$ in (24). In this case, the MMSE estimation problem can be solved analytically, and the solution satisfies the Riccati difference equation in (24).

3.6 Convergence of the robust Kalman filter

If the system state space matrices are time-invariant, it is well known that the recursion in the conventional Kalman filtering algorithm converges to a steady state estimator under the conditions listed in Proposition A.2. If system (3) is mean square stable, the robust Kalman filtering algorithm has a similar convergence property.

Theorem 3.3 *Consider the state estimation problem ($L = I$) with system (3), with all the state space matrices being time invariant and DD^T being positive definite⁶. If:*

1. *There exists $Q = Q^T$ with $Q > 0$ such that*

$$A^T Q A - Q + \sum_{j=1}^m (A_j^s)^T Q (A_j^s) < 0, \quad (26)$$

then the system (3) is mean square stable, and $\lim_{k \rightarrow \infty} X(k) = X$, where X is unique and independent of the initial condition $X(0) \in \mathbf{Co}\{X_1(0), \dots, X_p(0)\}$. Moreover, if the steady state matrices (A, \bar{B}, C, \bar{D}) , where

$$\bar{B} = \begin{bmatrix} A_1^s X & \dots & A_m^s X & B & B_1^s & \dots, & B_m^s \end{bmatrix}, \quad \bar{D} = \begin{bmatrix} C_1^s X & \dots & C_m^s X & D & D_1^s & \dots & D_m^s \end{bmatrix}, \quad (27)$$

satisfy the following condition:

2. *$(A - \bar{B}\bar{D}^T(\bar{D}\bar{D}^T)^{-1}C, \bar{B}(I - \bar{D}^T(\bar{D}\bar{D}^T)^{-1}\bar{D}))$ is stabilizable;*

or

$$\begin{bmatrix} A - e^{j\omega} I & \bar{B} \\ C & \bar{D} \end{bmatrix} \text{ has full row rank for all } \omega \in [0, 2\pi),$$

then the robust Kalman filter converges to a steady state linear time invariant estimator, i.e., for any $P(0) \geq 0$, we have $\lim_{k \rightarrow \infty} P^(k) = \bar{P}^* \geq 0$, $\lim_{k \rightarrow \infty} K^*(k) = \bar{K}^*$, $\lim_{k \rightarrow \infty} F^*(k) = \bar{F}^*$ and $(A + \bar{K}^*C)$ is stable, where $\{P^*(k), K^*(k), F^*(k)\}$ is the sequence consisting of the solutions to the robust Kalman filter.*

The proof of Theorem 3.3 is given in Appendix A.

⁶This condition can be relaxed to $DD^T + \sum_{i=1}^m (C_i^s X (C_i^s)^T + D_i^s (D_i^s)^T) > 0$ where $\lim_{k \rightarrow \infty} X(k) = X$.

4 A numerical example: equalizer for communication channels

We present an application of the robust Kalman filtering techniques proposed in this paper towards the design of equalizers for a communication channel. Consider the following system

$$\begin{aligned} \begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} &= \begin{bmatrix} 0.9 & 0.5 \\ 0 & 0.9 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \begin{bmatrix} 1 \\ 1 \end{bmatrix} s(k), \\ y(k) &= \begin{bmatrix} 1 + \zeta(k) & 1 + \zeta(k) \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + (5 + \zeta(k))s(k) + w(k). \end{aligned} \quad (28)$$

s is a signal that has a power of 0 dB and is transmitted through the channel. w is a white noise that corrupts the received signal y , with a power of -10 dB. The channel model (28) is affected by time-varying uncertainties ζ that are a combination of both deterministic and stochastic parametric uncertainties (see for example, [4]). The initial conditions of the state vectors, $x_1(0)$ and $x_2(0)$, are random variables and satisfy the second moment conditions

$$E[x_1(0)x_1(0)] \leq 1, \quad E[x_2(0)x_2(0)] \leq 1, \quad E[x_1(0)x_2(0)] = 0, \quad E[x_2(0)x_1(0)] = 0.$$

For the channel (28), we design an equalizer to estimate the input signal $s(k)$ (Figure 2).

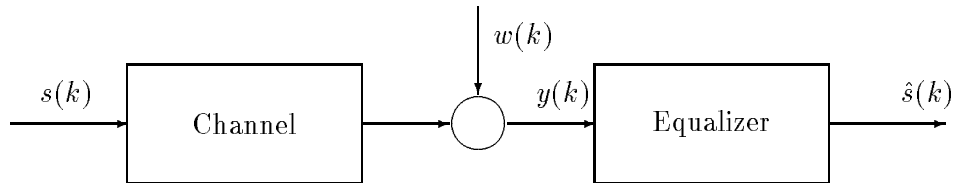


Figure 2: A communication channel with an equalizer.

We first add one more state variable in (28), so that the new model of the channel is now

$$\begin{aligned} \begin{bmatrix} x_1(k+1) \\ x_2(k+1) \\ x_3(k+1) \end{bmatrix} &= \begin{bmatrix} 0.9 & 0.5 & 1 \\ 0 & 0.9 & 1 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \\ x_3(k) \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} s(k+1), \\ y(k) &= \begin{bmatrix} 1 + \zeta(k) & 1 + \zeta(k) & 5 + \zeta(k) \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \\ x_3(k) \end{bmatrix} + w(k), \\ z(k) &= \begin{bmatrix} 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \\ x_3(k) \end{bmatrix}. \end{aligned} \quad (29)$$

Assuming a zero mean white noise model for the input signal s , we design an equalizer using the robust Kalman filtering techniques developed in this paper. The first case considered is when the channel is on its nominal condition and does not have any uncertainty, i.e., $\zeta(k) = 0$; this is an ideal channel, and the corresponding equalizer is a conventional Kalman filter. In Figure 3, we compare a Kalman filter initialized optimally by solving (22), with Kalman filters initialized

using an *ad hoc* scheme [2] ($P(0) = \eta I$, where $\eta = 5, 10$ and 20). The Mean Square Error (MSE) estimates are obtained by averaging 200 runs. Under the condition of $\zeta(k) = 0$ in channel (29), it can be easily checked that the Kalman filter for this system converges (see the convergence condition in Proposition A.2). Therefore after 10 steps, we can see that the MSE of Kalman filters with different initializations are close to each other and converge to a constant. However for the transient performance (less than 10 steps), the *ad hoc* initialized Kalman filter is significantly worse than that of the optimally initialized Kalman filter. This example indicates that the optimal initialization by solving (22) can be superior to an *ad hoc* initialization, and can improve the transient performance of conventional Kalman filters.

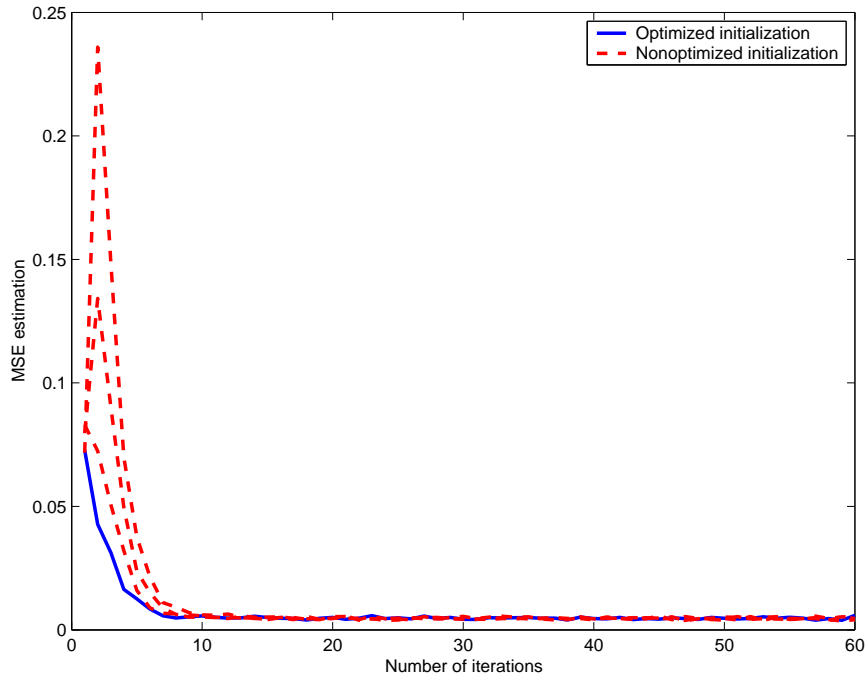


Figure 3: Illustration of the performance improvement of the conventional Kalman filter with optimal initialization.

In the second example, we compare the performance of the conventional Kalman filtering algorithm with the robust Kalman filtering algorithm in the presence of uncertainties. Specifically, the time varying uncertainties satisfy $\zeta(k) = 0.1\zeta_d(k) + 0.5\zeta_s(k)$, where: $\zeta_d(k)$ is deterministic, and $|\zeta_d(k)| < 1$ for all k , and can be measured in real-time; ζ_s is a zero-mean white noise process with a power of 0 dB, and is independent of w and s . Figure 4 shows a comparison of the experimentally obtained mean square error values, averaged over 200 runs, obtained with the conventional Kalman filters, and the robust Kalman filter. For a fair comparison, we include the simulation results of both the conventional Kalman filter initialized using an *ad hoc* scheme, and the conventional Kalman filter initialized optimally by solving (22). For the nonoptimally initialized conventional Kalman filter, the performance is significantly worse than that of the other two filtering algorithms which are initialized optimally. For the optimally initialized conventional Kalman filter, since it has

a similar initialization as the robust Kalman filter, its transient performance at the beginning of the recursion (less than 10 steps) is similar to that of the robust Kalman filter. However, since the conventional Kalman filter does not consider the uncertainties in system matrices, its performance degrades significantly thereafter. With the robust Kalman filtering algorithm, an upper bound of the mean square estimation error over all possible uncertainties is minimized recursively, and therefore the performance is much improved. It can be seen from Figure 4 that after 10 steps of recursions, the robust Kalman filter yields a much lower mean-squared error than that with the conventional Kalman filters.

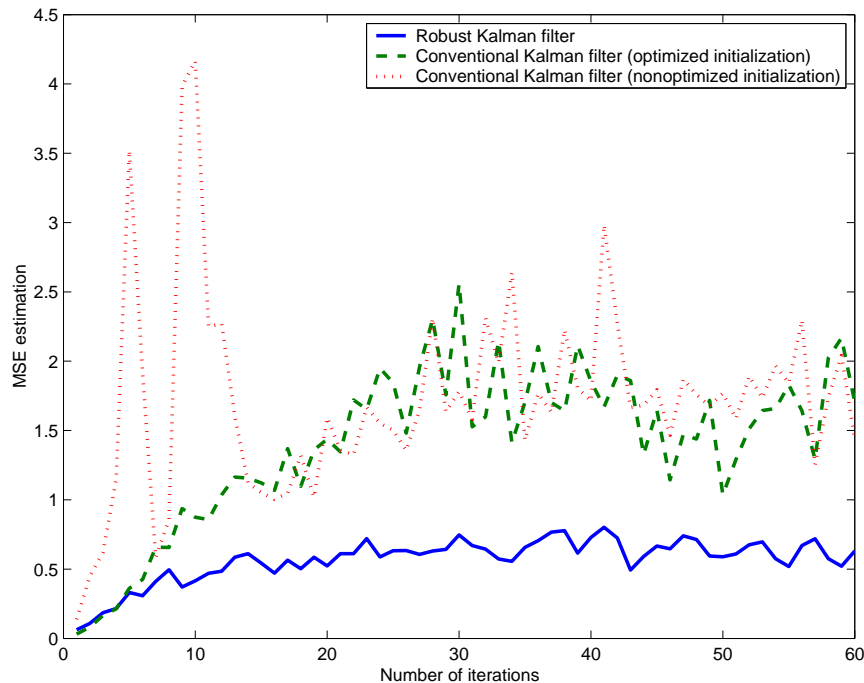


Figure 4: Performance comparison between the robust and conventional Kalman filters.

5 Conclusion

We have developed a robust Kalman filtering algorithm for linear time-varying systems with stochastic parametric uncertainties. We have shown that for systems without uncertainties, the robust Kalman filter reduces to the conventional Kalman filter; for systems with stochastic parametric uncertainties, it offers significant improvement in performance. If the system is mean square stable and the state space matrices are time-invariant, the robust Kalman filter converges to a steady state estimator. Our filtering algorithm is formulated as a convex optimization problem with linear matrix inequality constraints, and can be implemented numerically efficiently. We have established that the techniques presented in the paper can be used to optimally initialize other recursive algorithms, including the conventional Kalman filtering algorithm, with an attendant improvement in transient performance. Finally, we have shown via a numerical example that the techniques de-

veloped in this paper can be applied towards the design of equalizers for communication channels, with much improved results.

A Appendix

Lemma A.1 ([1, page 131]) *The system*

$$x(k+1) = A_{\Delta}(k)x(k), \quad (30)$$

where $x(k) \in \mathbf{R}^n$,

$$A_{\Delta}(k) = A + \sum_{i=1}^m A_i^s \zeta_i^s(k),$$

and $\zeta_i^s, i = 1, \dots, m$ are zero-mean random processes with $E[\zeta_i^s(k)\zeta_j^s(l)] = \delta(i-j)\delta(k-l)$ is mean square stable if and only if there exists a matrix $Q = Q^T \in \mathbf{R}^{n \times n}$ with $Q > 0$ such that

$$A^T Q A - Q + \sum_{i=1}^m (A_i^s)^T Q (A_i^s) < 0. \quad (31)$$

Moreover, the quadratic Lyapunov function $V(x(k)) = E[x(k)^T Q x(k)]$ is a monotonically decreasing function of k , with $\lim_{k \rightarrow \infty} V(x(k)) = 0$.

Note that the correlation of the state in system (30) satisfies the recursion

$$X(k+1) = AX(k)A^T + \sum_{j=1}^m A_j^s X(k)(A_j^s)^T,$$

where $X(k) = E[x(k)x(k)^T]$ and $X(0)$ is the initial condition. Lemma A.1 implies that condition (31) is sufficient and necessary to have $\lim_{k \rightarrow \infty} X(k) = 0$.

The following result about the convergence of the conventional Kalman filtering algorithm is well-known. More details about the Proposition A.2 can be found in the literature; see for example [8].

Proposition A.2 *Let $A \in \mathbf{R}^{n \times n}$, $B \in \mathbf{R}^{n \times n_w}$, $C \in \mathbf{R}^{n_y \times n}$, and $D \in \mathbf{R}^{n_y \times n_w}$. Suppose $DD^T > 0$. If one of the following conditions holds,*

1. (A, C) is detectable and $(A - BD^T(DD^T)^{-1}C, B(I - D^T(DD^T)^{-1}D))$ is stabilizable;
2. (A, C) is detectable and

$$\begin{bmatrix} A - e^{j\omega}I & B \\ C & D \end{bmatrix}$$

has full row rank for all $\omega \in [0, 2\pi)$;

then for any initial condition $P(0) \geq 0$, the sequence of solutions $\{P^*(k), K^*(k)\}$ to the recursion (24) converges to a unique steady state solution

$$\lim_{k \rightarrow \infty} P^*(k) = P^* \quad \text{and} \quad \lim_{k \rightarrow \infty} K^*(k) = K^*.$$

Furthermore, $P^* \geq 0$, $K^* = -(AP^*C^T + BD^T)(CP^*C^T + DD^T)^{-1}$ and $(A + K^*C)$ is stable.

Proposition A.3 Let $\{X(k)\}$ be the sequence consisting of the solutions to the recursions

$$X(k+1) = AX(k)A^T + BB^T + \sum_{j=1}^m \left(A_j^s X(k) (A_j^s)^T + B_j^s (B_j^s)^T \right), \quad k = 1, 2, \dots, \quad (32)$$

with initial condition $X(0) \geq 0$, where A , B , A_j^s , B_j^s are constant matrices. The sequence $\{X(k)\}$ is convergent and the limit is independent of $X(0)$ if and only if there exists some matrix $Q > 0$ such that

$$A^T Q A - Q + \sum_{j=1}^m (A_j^s)^T Q (A_j^s) < 0, \quad (33)$$

i.e., the system is mean square stable.

Proof: We first show that the sequence $\{X(k)\}$ obtained from (32) is convergent and the limit is independent of $X(0)$ if and only if the sequence $\{\hat{X}(k)\}$ obtained from

$$\hat{X}(k+1) = A\hat{X}(k)A^T + \sum_{j=1}^m A_j^s \hat{X}(k) (A_j^s)^T \quad (34)$$

is convergent and $\lim_{k \rightarrow \infty} \hat{X}(k) = 0$.

Suppose the limit of the sequence $\{\hat{X}(k)\}$ exists and $\lim_{k \rightarrow \infty} \hat{X}(k) = 0$. For any $X(0)$, define $\hat{X}(k) = X(k+1) - X(k)$. $\hat{X}(k)$ satisfies the recursion (34) and $\lim_{k \rightarrow \infty} \hat{X}(k) = 0$. Then it can be shown that $\{X(k)\}$ is convergent.

To show that the limit of $\{X(k)\}$ is unique, suppose that $\{X_1(k)\}$ and $\{X_2(k)\}$ are two sequences of solutions to (32), corresponding to two different initial conditions $X_1(0)$ and $X_2(0)$, with $\lim_{k \rightarrow \infty} X_1(k) = X_1$ and $\lim_{k \rightarrow \infty} X_2(k) = X_2$. Then, it is easily verified that with $\hat{X}(k) = X_1(k) - X_2(k)$, the sequence $\{\hat{X}(k)\}$ consists of solutions of the recursive equation (34); then $\lim_{k \rightarrow \infty} \hat{X}(k) = 0$ implies $X_1 = X_2$.

Now suppose the limit of the sequence $\{X(k)\}$ exists and is independent of $X(0)$. We show that $\lim_{k \rightarrow \infty} \hat{X}(k) = 0$.

For any $\hat{X}(0)$, let $X_1(0) = \hat{X}(0)$, $X_2(0) = 0$ and $\{X_i(k)\}$, $i = 1, 2$ be the corresponding sequences from recursion (32). By recursion (34), we have $\hat{X}(k) = X_1(k) - X_2(k)$. Since $\lim_{k \rightarrow \infty} X_1(k) = \lim_{k \rightarrow \infty} X_2(k)$, we get $\lim_{k \rightarrow \infty} \hat{X}(k) = 0$.

Finally by Lemma A.1, the condition (33) is necessary and sufficient for $\{\hat{X}(k)\}$ to converge to zero. This completes the proof. \square

Proof of Theorem 3.2:

Starting with the sequence $\{P^*(k), K^*(k)\}$ consisting of the solutions to the optimization problem (24), it is clear that

$$\tilde{P}^*(k) = P^*(k) \quad \text{and} \quad \tilde{K}^*(k) = K^*(k), \quad k = 1, 2, \dots,$$

must be (possibly nonunique) optimizers for problem (25) (this follows immediately from (25d)). Thus, it remains to show that $\tilde{P}^*(k) = P^*(k)$ and $\tilde{K}^*(k) = K^*(k)$ are the only candidates as optimizers for problem (25). In other words, it suffices to show that for a given $\tilde{P}^*(k-1)$, the

optimal solutions $\tilde{P}^*(k)$ and $\tilde{K}^*(k)$ in (25) are unique. However, note that given $\tilde{P}^*(k-1)$ and $\tilde{K}^*(k-1)$, $\tilde{P}^*(k)$ is given simply as $f(\tilde{P}^*(k-1), \tilde{K}^*(k-1))$, thus we only need to show that given $\tilde{P}^*(k-1)$, $\tilde{K}^*(k-1)$ that solves (25) is unique.

Now, let $\{\tilde{P}^*(k), \tilde{K}^*(k), \tilde{F}^*(k)\}$ be the sequence consisting of the solutions to the optimization problem (25). Assuming an optimal value for $\tilde{P}^*(k-1)$, we consider the optimization problem (25). For convenience, we introduce new notation for $M(k)$:

$$g(K(k-1), F(k)) = (I + F(k)C)f(P(k-1), K(k-1))(I + F(k)C)^T + F(k)DD^T F(k)^T.$$

We also denote

$$F_{\text{opt}}(K(k-1)) = \underset{F(k)}{\text{argmin}} \mathbf{Tr}(g(K(k-1), F(k))).$$

Note that $\tilde{F}^*(k) = F_{\text{opt}}(\tilde{K}^*(k-1))$.

Now for any $P(k) = f(P(k-1), K(k-1)) \geq 0$, since $(CP(k)C^T + DD^T) > 0$, it can be easily verified that if $F(k) \neq F_{\text{opt}}(K(k-1))$, then,

$$g(K(k-1), F(k)) \not\geq g(K(k-1), F_{\text{opt}}(K(k-1))).$$

Therefore $\mathbf{Tr}(g(K(k-1), F(k))) > \mathbf{Tr}(g(K(k-1), F_{\text{opt}}(K(k-1))))$. Thus, $F_{\text{opt}}(K(k-1))$ is uniquely determined from $K(k-1)$.

To show that $\tilde{K}^*(k-1)$ is unique, it suffices to show that for any $K(k-1) \neq \tilde{K}^*(k-1)$, $\mathbf{Tr}(g(K(k-1), F_{\text{opt}}(K(k-1)))) > \mathbf{Tr}(g(\tilde{K}^*(k-1), F_{\text{opt}}(\tilde{K}^*(k-1))))$. Let $P_1(k) = f(\tilde{P}^*(k-1), K(k-1))$ and $P_2(k) = f(\tilde{P}^*(k-1), \tilde{K}^*(k-1))$. Then we have $P_1(k) \not\geq P_2(k)$. We now consider two cases:

1. If $F_{\text{opt}}(K(k-1)) \neq F_{\text{opt}}(\tilde{K}^*(k-1))$, then

$$g(K(k-1), F_{\text{opt}}(K(k-1))) \not\geq g(\tilde{K}^*(k-1), F_{\text{opt}}(\tilde{K}^*(k-1)))$$

Therefore $\mathbf{Tr}(g(K(k-1), F_{\text{opt}}(K(k-1)))) > \mathbf{Tr}(g(\tilde{K}^*(k-1), F_{\text{opt}}(\tilde{K}^*(k-1))))$.

2. If $F_{\text{opt}}(K(k-1)) = F_{\text{opt}}(\tilde{K}^*(k-1))$, then

$$P_1(k)C^T(CP_1(k)C^T + DD^T)^{-1} = P_2(k)C^T(CP_2(k)C^T + DD^T)^{-1}.$$

After simple manipulations, we get

$$DD^T(CP_1(k)C^T + DD^T)^{-1} = DD^T(CP_2(k)C^T + DD^T)^{-1}.$$

Since $DD^T > 0$ and $P_1(K) \geq P_2(K)$, we have $C(P_1(k) - P_2(k)) = 0$. By (25d), this implies that $\mathbf{Tr}(g(K(k-1), F_{\text{opt}}(K(k-1)))) > \mathbf{Tr}(g(\tilde{K}^*(k-1), F_{\text{opt}}(\tilde{K}^*(k-1))))$.

This completes the proof. □

Now we are ready to prove Theorem 3.3, that the robust Kalman filtering algorithm is convergent.

Proof of Theorem 3.3:

With all the state space matrices in system (3) being time-invariant, we have $A(k) = A$, $B(k) = B$ and

$$\begin{aligned}\hat{B}(k) &= \begin{bmatrix} A_1^s X(k)^{1/2} & \dots & A_m^s X(k)^{1/2} & B & B_1^s & \dots & B_m^s \end{bmatrix}, \\ \hat{D}(k) &= \begin{bmatrix} C_1^s X(k)^{1/2} & \dots & C_m^s X(k)^{1/2} & D & D_1^s & \dots & D_m^s \end{bmatrix}.\end{aligned}$$

The correlation of the state $X(k) = E[x(k)x(k)^T]$ satisfies the recursion

$$X(k+1) = AX(k)A^T + BB^T + \sum_{j=1}^m \left(A_j^s X(k)(A_j^s)^T + B_j^s (B_j^s)^T \right), \quad k = 0, 1, \dots$$

The mean square stability of system (3) follows directly from Lemma A.1.

By Proposition A.3, we have $\lim_{k \rightarrow \infty} X(k) = X$, where X is unique and independent of the initial condition $X(0) \in \mathbf{Co}\{X_1(0), \dots, X_p(0)\}$.

With $\lim_{k \rightarrow \infty} \hat{B}(k) = \bar{B}$ and $\lim_{k \rightarrow \infty} \hat{D}(k) = \bar{D}$, where \bar{B} and \bar{D} are defined in (27), we now define another set of recursions

$$\min_{\bar{P}(k), \bar{M}(k), \bar{K}(k-1), \bar{F}(k)} \mathbf{Tr}(\bar{M}(k)), \quad (35a)$$

$$\begin{aligned}\bar{f}(\bar{P}(k-1), \bar{K}(k-1)) &\triangleq (A + \bar{K}(k-1)C)\bar{P}(k-1)(A + \bar{K}(k-1)C)^T \\ &\quad + (\bar{K}(k-1)\bar{D} + \bar{B})(\bar{K}(k-1)\bar{D} + \bar{B})^T, \end{aligned} \quad (35b)$$

$$\bar{P}(k) = \bar{f}(\bar{P}(k-1), \bar{K}(k-1)), \quad (35c)$$

$$\bar{M}(k) = (I + \bar{F}(k)C)\bar{P}(k)(I + \bar{F}(k)C)^T + \bar{F}(k)\bar{D}\bar{D}^T\bar{F}(k)^T. \quad (35d)$$

Let $\bar{K}^*(k)$ and $\bar{F}^*(k)$ denote the optimal Kalman gains of (35). Let $K^*(k)$ and $F^*(k)$ denote the optimal Kalman gains of the robust Kalman filter (15). From Theorem 3.2, equation (35) has the same unique optimal solution of $\bar{P}^*(k)$ and $\bar{K}^*(k)$ as that of equation (24), with B replaced by \bar{B} and D replaced by \bar{D} . From Proposition A.2, we have $\lim_{k \rightarrow \infty} \bar{K}^*(k) = \bar{K}^*$ and $(A + \bar{K}^*C)$ is stable. Thus

$$\left\| \sum_{j=1}^i \prod_{k=j}^i (A + \bar{K}^*(k)C)(A + \bar{K}^*(k)C)^T \right\| \leq T_1(\bar{P}(0)),$$

where $T_1(\bar{P}(0))$ is a uniform bound over $i > 0$, and depends only on $\bar{P}(0)$.

From (10) and (35b), it follows that for any small number ϵ , there exists N , such that whenever $k \geq N$, we have

$$\|f(P(k), K^*(k)) - \bar{f}(P(k), \hat{K}^*(k))\| \leq T_2(P(k))\epsilon,$$

where $\hat{K}^*(k) = \operatorname{argmin}_{K(k)} \bar{f}(P(k), K(k))$, $T_2(P(k))$ is a constant that depends on $P(k)$. $T_2(P(k))$ is finite if $P(k)$ is bounded.

By setting $P(N) = \bar{P}(N)$, we get $\|P^*(N+1) - \bar{P}^*(N+1)\| \leq T_2(P(N))\epsilon$, where $P^*(N+1)$ and $\bar{P}^*(N+1)$ are optimal solutions to recursions (15) and (35) respectively. Next, we have

$$\begin{aligned} \|P^*(N+2) - \bar{P}^*(N+2)\| &= \|f(P^*(N+1), K^*(N+1)) - \bar{f}(\bar{P}^*(N+1), \bar{K}^*(N+1))\| \\ &\leq \left(T_2(P(N)) \|(A + \bar{K}^*(N+1)C)(A + \bar{K}^*(N+1)C)^T\| + T_2(P^*(N+1)) \right) \epsilon. \end{aligned}$$

Recursively for $i = 2, 3, \dots$, we get

$$\begin{aligned} \|P^*(N+i) - \bar{P}^*(N+i)\| &= \|f(P^*(N+i-1), K^*(N+i-1)) - \bar{f}(\bar{P}^*(N+i-1), \bar{K}^*(N+i-1))\| \\ &\leq \left(T_2(P^*(N+i-1)) + \left\| \sum_{j=1}^{i-1} T_2(P^*(N+j-1)) \prod_{k=N+j}^{N+i-1} (A + \bar{K}^*(k)C)(A + \bar{K}^*(k)C)^T \right\| \right) \epsilon \\ &\leq \left(T_2(P^*(N+i-1)) + \max_{j=0, \dots, i-2} T_2(P^*(N+j)) T_1(P(N)) \right) \epsilon. \end{aligned}$$

Using standard arguments, we obtain that $\{P^*(k)\}$ is also bounded, with the bound depending only on $P(0)$. Therefore, there exist finite constants $T_3(P(0))$ and $T_4(P(0))$, that depend on $P(0)$, such that for any $k \geq 0$, $T_1(P^*(k)) \leq T_3(P(0))$ and $T_2(P^*(k)) \leq T_4(P(0))$. Thus, we conclude that $\|P^*(N+i) - \bar{P}^*(N+i)\|$ is bounded and that

$$\|P^*(N+i) - \bar{P}^*(N+i)\| \leq [T_4(P(0))T_3(P(0)) + T_4(P(0))]\epsilon.$$

Since ϵ can be made arbitrarily small by choosing N to be large enough, we obtain

$$\lim_{k \rightarrow \infty} P^*(k) = \lim_{k \rightarrow \infty} \bar{P}^*(k) = \bar{P}^*.$$

Similar arguments establish that $\lim_{k \rightarrow \infty} K^*(k) = \lim_{k \rightarrow \infty} \bar{K}^*(k) = \bar{K}^*$ and that $\lim_{k \rightarrow \infty} F^*(k) = \lim_{k \rightarrow \infty} \bar{F}^*(k) = \bar{F}^*$. \square

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